



# Control Systems in the presence of Computational Problems

Martina Maggio

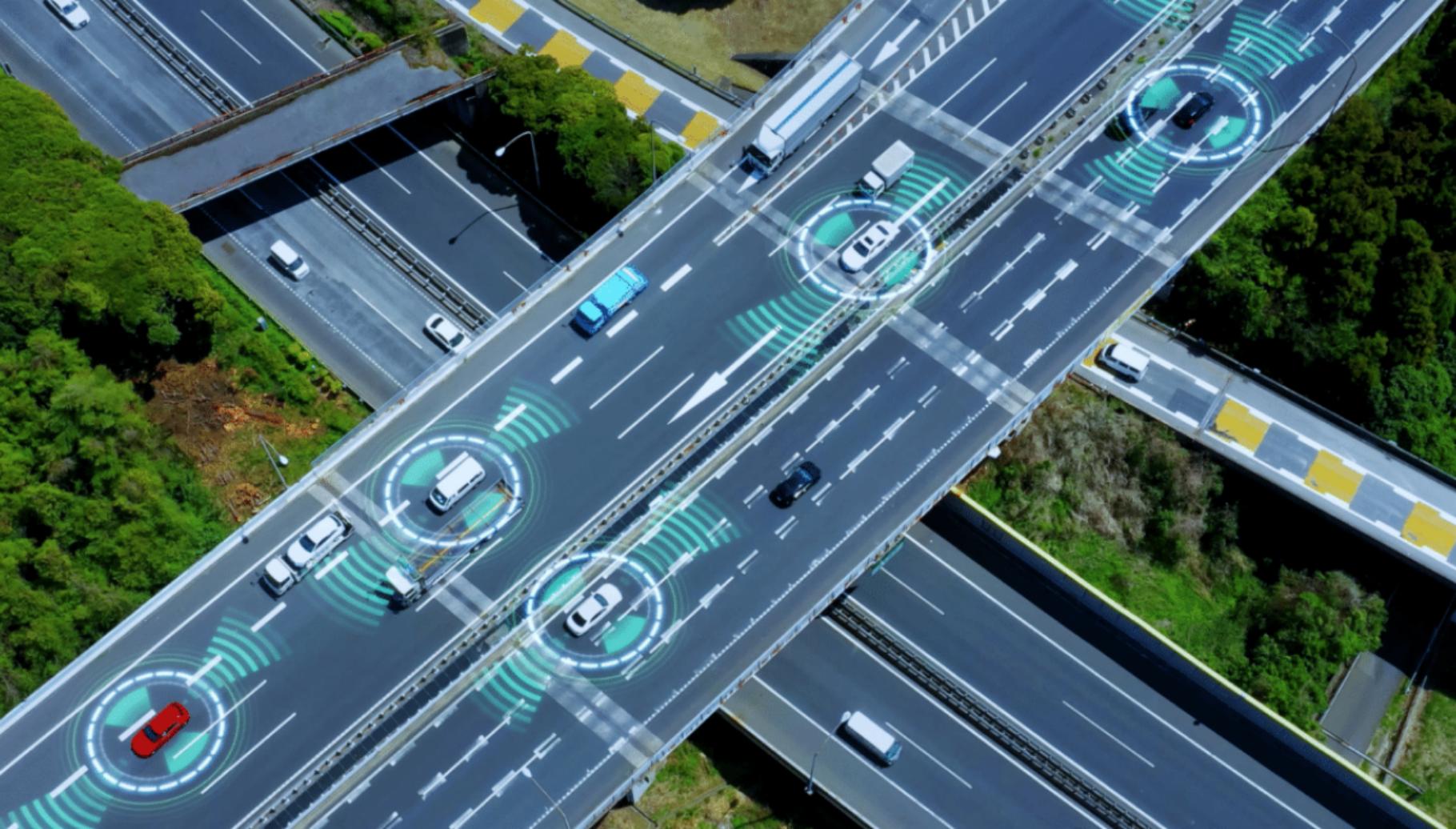
## Control: The Hidden Technology<sup>1</sup>

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<sup>1</sup>This is the title of a famous *lectio magistralis* on control given by Karl-Johan Åström.







## In a nutshell...

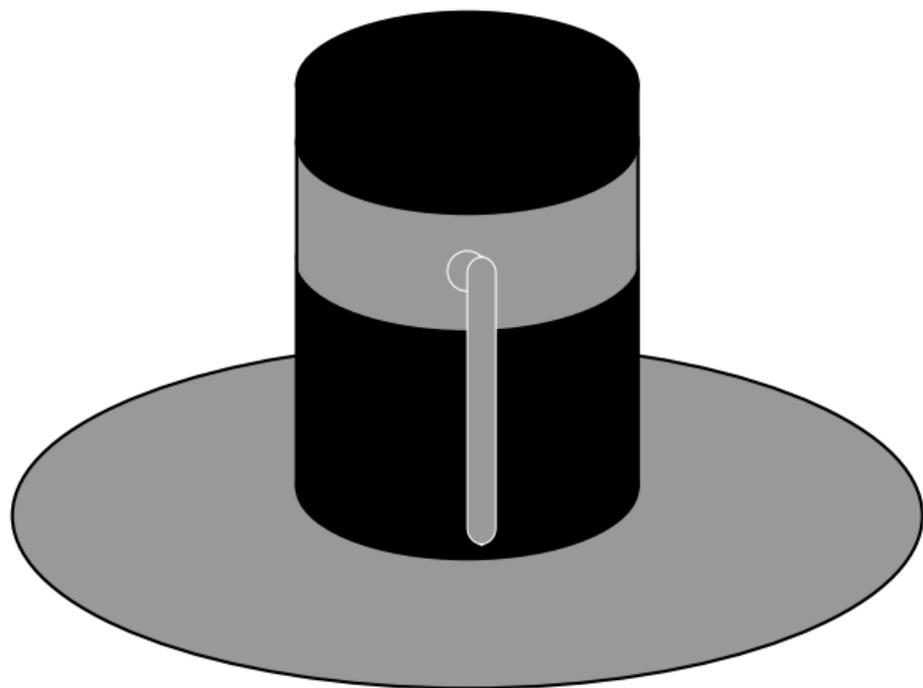
- ▶ Controllers are software programs that run on hardware
- ▶ As such, they can experience computational problems
- ▶ For the rest of this talk: faults causes *deadline misses*

## In a nutshell...

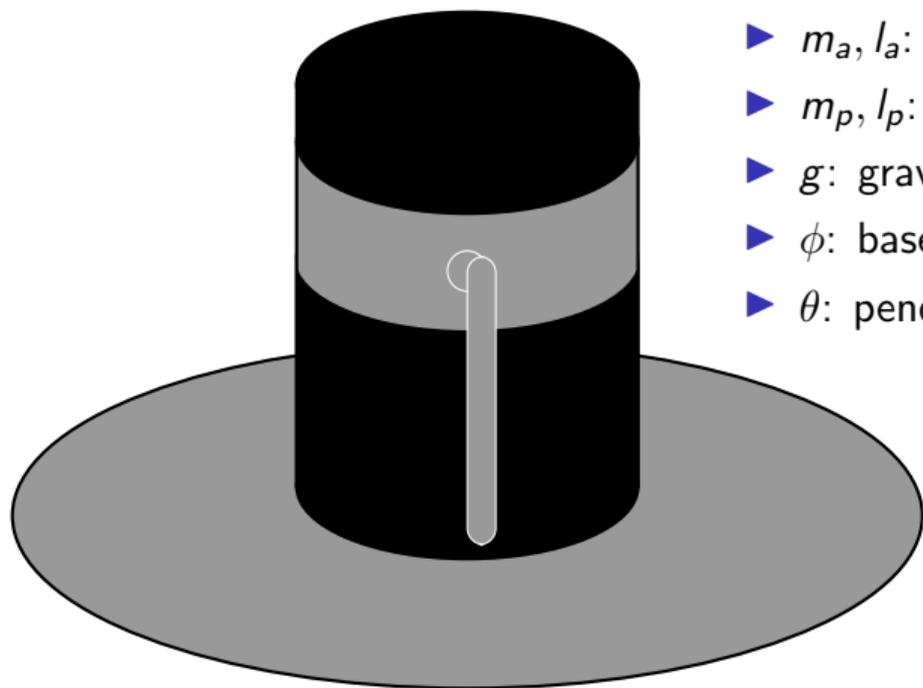
- ▶ Controllers are software programs that run on hardware
- ▶ As such, they can experience computational problems
- ▶ For the rest of this talk: faults causes *deadline misses*
- ▶ If we run these controller *in practice* we see that very often deadline misses are not a problem – but: can we *certify* that the system “will not misbehave” despite the presence of deadline misses?

# Control Design

## Modelling the Physical Phenomena<sup>2</sup>

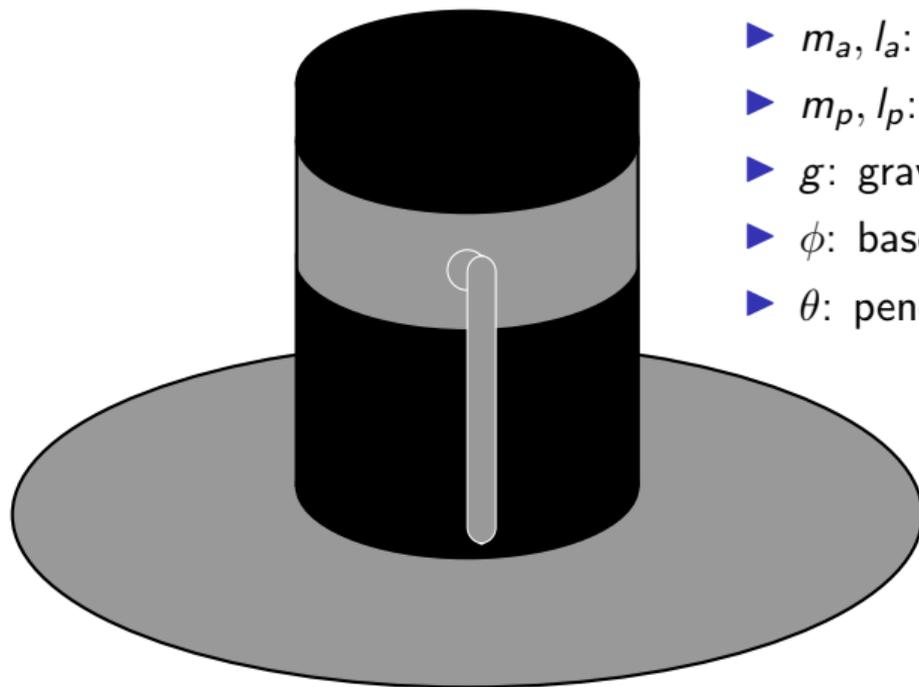


## Modelling the Physical Phenomena<sup>2</sup>



- ▶  $J$ : moment of inertia of the center pillar
- ▶  $m_a, l_a$ : mass and length of first arm
- ▶  $m_p, l_p$ : mass and length of pendulum arm
- ▶  $g$ : gravitational acceleration constant
- ▶  $\phi$ : base angle
- ▶  $\theta$ : pendulum angle

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- ▶  $\phi$ : base angle
- ▶  $\theta$ : pendulum angle

- ▶  $\alpha := J + (\frac{1}{3}m_a + m_p) l_a^2$
- ▶  $\beta := \frac{1}{3}m_p l_p^2$
- ▶  $\gamma := \frac{1}{2}m_p l_a l_p$
- ▶  $\delta := \frac{1}{2}m_p l_p g$

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$$\frac{d\phi}{dt} = \dot{\phi}$$

$$\frac{d\dot{\phi}}{dt} = \frac{1}{\alpha\beta - \gamma^2 + (\beta^2 + \gamma^2) \sin^2 \theta} \{ \beta\gamma(\sin^2 \theta - 1) \sin \theta \dot{\phi}^2 + \dots \}$$

$$\frac{d\theta}{dt} = \dot{\theta}$$

$$\frac{d\dot{\theta}}{dt} = \frac{1}{\alpha\beta - \gamma^2 + (\beta^2 + \gamma^2) \sin^2 \theta} \{ \beta(\alpha + \beta \sin^2 \theta) \cos \theta \sin \theta \dot{\phi}^2 + \dots \}$$

$$\alpha := J + \left(\frac{1}{3}m_a + m_p\right) l_a^2$$

$$\beta := \frac{1}{3}m_p l_p^2$$

$$\gamma := \frac{1}{2}m_p l_a l_p$$

$$\delta := \frac{1}{2}m_p l_p g$$

<sup>2</sup>For the full derivation, see Magnus Gäfvert, Modelling the Furuta Pendulum, ISSN 0280-5316

# Modelling the Physical Phenomena

- ▶ Identifying system state, input, and output
- ▶ Non-linear resulting model
- ▶ Determining the system *equilibria* and linearizing the model

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$$y(t) = C_c x(t) + D_c u(t)$$

# Modelling the Physical Phenomena

- ▶ Identifying system state, input, and output
- ▶ Non-linear resulting model
- ▶ Determining the system *equilibria* and linearizing the model
- ▶ Discretizing with time step  $T$

$$x_{k+1} = A_d x_k + B_d u_k$$

$$y_k = C_d x_k + D_d u_k$$

## Example: Furuta Pendulum model

$$x_{k+1} = A_d x_k + B_d u_k$$

$$y_k = C_d x_k + D_d u_k$$

- ▶  $x = [\theta \quad \dot{\theta} \quad \dot{\phi}]^T$ ,  $y = x$ ,  $T = 5 \text{ ms}$
- ▶  $u$  is the torque applied at the base level

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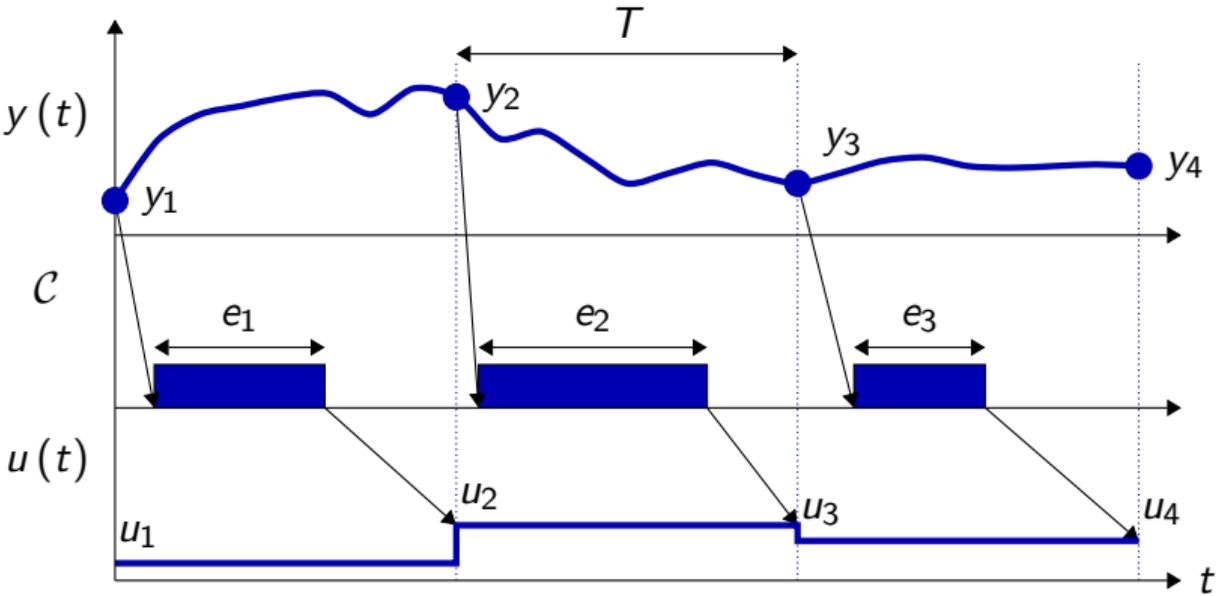
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- ▶  $x = [\theta \ \dot{\theta} \ \dot{\phi}]^T$ ,  $y = x$ ,  $T = 5 \text{ ms}$
- ▶  $u$  is the torque applied at the base level

Around the *upright* equilibrium point:

$$A_d = \begin{bmatrix} 1.001 & 0.005 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}, B_d = \begin{bmatrix} -0.083 \\ -33.2 \\ 38.6 \end{bmatrix}, C_d = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}, D_d = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}.$$

# Controller Nominal Execution



## Synthesizing the Controller

- ▶ Based on objectives (like speed of convergence and ability to reject disturbances) we can pick a control algorithm (which executes periodically inside  $e_k$ )
  - many alternatives: state/output feedback, PID, LQR, LQG, MPC, ...
- ▶ and verify that the closed-loop behaves in the desired way.

## Example: Furuta Pendulum control synthesis

$$x_{k+1} = A_d x_k + B_d u_k$$

$$u_{k+1} = K y_k = K x_k = [0.375 \quad 0.025 \quad 0.0125] x_k$$

- ▶ Output feedback controller (but  $y = x$ , hence state feedback)
- ▶ At the beginning of every iteration we sense  $y$ , and calculate the next  $u$
- ▶ Autonomous behavior:  $x_{k+1} = A_d x_k + B_d K x_{k-1}$

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$$\tilde{x}_k = \begin{bmatrix} x_k \\ x_{k-1} \end{bmatrix}, \quad \tilde{x}_{k+1} = \begin{bmatrix} x_{k+1} \\ x_k \end{bmatrix} = \underbrace{\begin{bmatrix} A_d & B_d K \\ I & 0 \end{bmatrix}}_A \begin{bmatrix} x_k \\ x_{k-1} \end{bmatrix} = A \tilde{x}_k$$

## Verifying the Control Design

- ▶ Typical assumptions in terms of computation:
  - instantaneous sensing and actuation
  - instantaneous computation
  - no communication overhead
- ▶ The design framework that we used is already employing a *one-step delay* paradigm, to take advantage of *predictable communication and execution times*

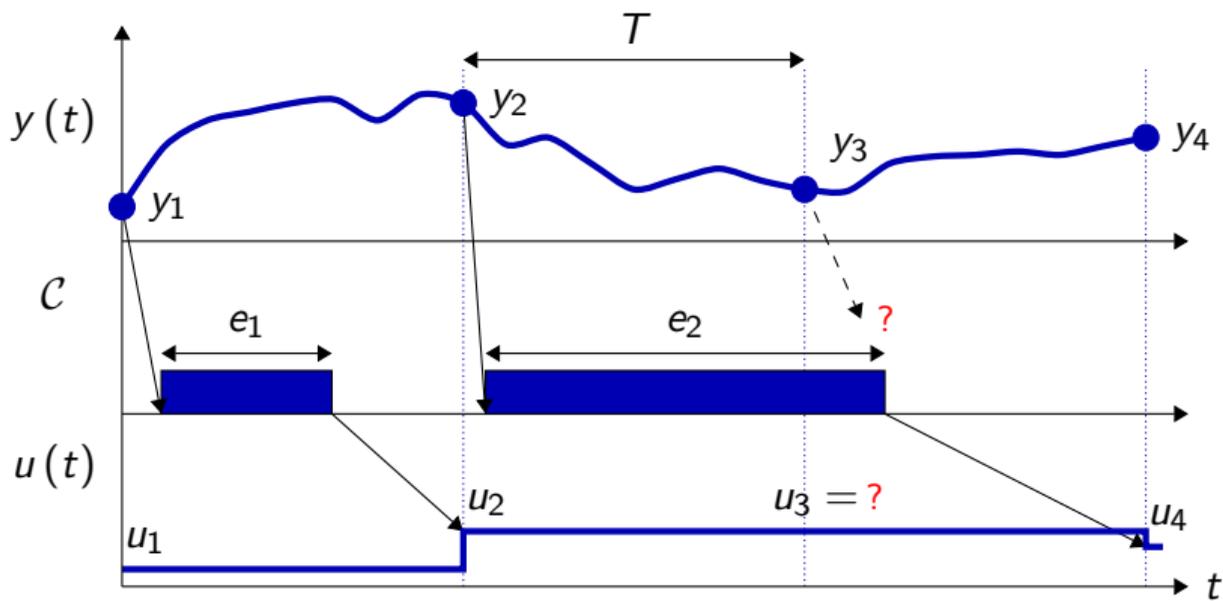
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- ▶ The design framework that we used is already employing a *one-step delay* paradigm, to take advantage of *predictable communication and execution times*
- ▶ If the spectral radius  $\rho(A)$  is less than 1, the closed-loop system is stable,

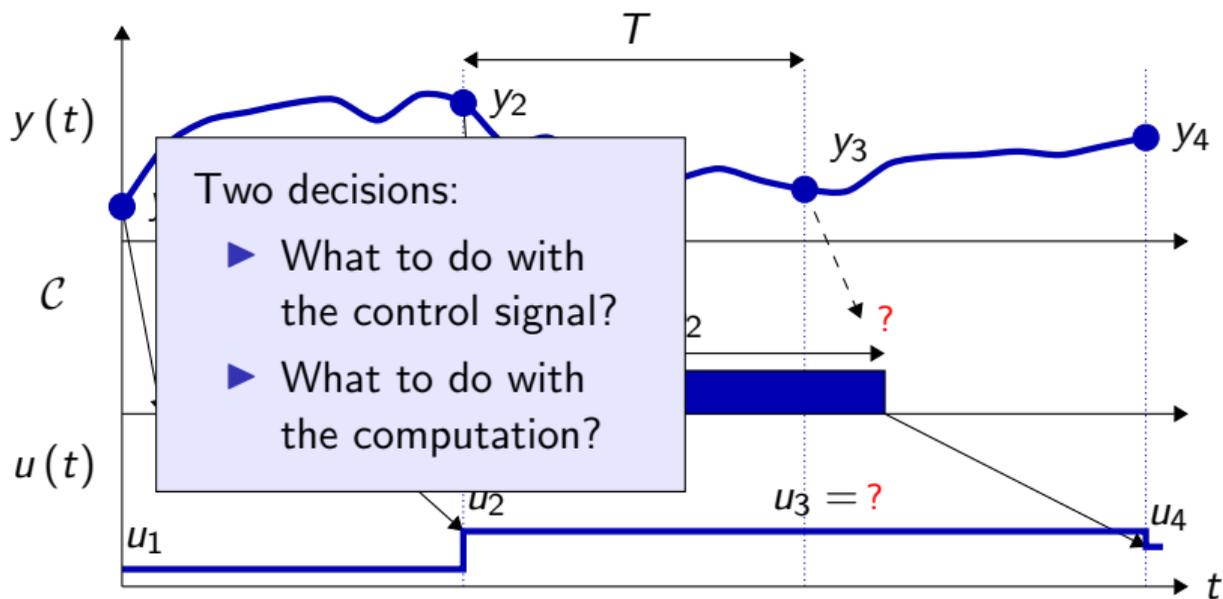
$$\rho(A) = \max |\lambda(A)|$$

What if there are deadline misses?

# Missing a Deadline



# Missing a Deadline



## Missing a Deadline

For the control signal<sup>3</sup>

- ▶ **Hold**: keeping the previous value
- ▶ **Zero**: set the control signal to zero

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<sup>3</sup>Steffen Linsenmayer and Frank Allgöwer, CDC 2017

“Stabilization of networked control systems with weakly hard real-time dropout description”

## Missing a Deadline

For the control signal<sup>3</sup>

- ▶ **Hold:** keeping the previous value
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For the computation<sup>4</sup>

- ▶ **Kill:** kill the current task with a clean reset, nothing happened
- ▶ **Skip-Next:** let the current task continue but do not start a new one in the next period and wait for the following activation

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<sup>3</sup>Steffen Linsenmayer and Frank Allgöwer, CDC 2017

“Stabilization of networked control systems with weakly hard real-time dropout description”

<sup>4</sup>Anton Cervin, IFAC World Congress 2005

“Analysis of overrun strategies in periodic control tasks.”

Kill&Zero

# Kill&Zero

## Hit

$$x_{k+1} = A_d x_k + B_d u_k$$

$$u_{k+1} = K x_k$$

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$$\begin{bmatrix} x_{k+1} \\ u_{k+1} \end{bmatrix} = \underbrace{\begin{bmatrix} A_d & B_d \\ K & 0 \end{bmatrix}}_{A_H} \begin{bmatrix} x_k \\ u_k \end{bmatrix}$$

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We don't have much hope  
to guarantee stability...  
...unless we add constraints!

## Constraint Example

“We cannot miss more than  $n$  consecutive deadlines”<sup>5</sup>

...means that the system switches arbitrarily between matrices in  $\Sigma$ :

$$\Sigma = \{A_H A_M^i \mid i \in \mathbb{Z}, 0 \leq i \leq n\}$$

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<sup>5</sup>Martina Maggio, Arne Hamann, Eckart Mayer-John, Dirk Ziegenbein, ECRTS 2020  
“Control System Stability under Consecutive Deadline Misses Constraints”

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corresponds to  $i$  misses  
followed by 1 hit

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<sup>5</sup>Martina Maggio, Arne Hamann, Eckart Mayer-John, Dirk Ziegenbein, ECRTS 2020  
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# Joint Spectral Radius

We can use a result<sup>6</sup> on switching systems, that states that the system that arbitrarily switches among matrices in  $\Sigma$  is asymptotically stable if and only if the joint spectral radius<sup>7</sup>  $\rho(\Sigma)$  is less than 1

$$\begin{aligned}\rho_\mu(\Sigma) &= \sup \{ \rho(A)^{\frac{1}{\mu}} : A \in \Sigma^\mu \} \\ \rho(\Sigma) &= \limsup_{\mu \rightarrow \infty} \rho_\mu(\Sigma)\end{aligned}$$

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<sup>6</sup>Gian-Carlo Rota and Gilbert Strang, *Indagationes Mathematicae*, 63:379-381, 1960  
“A note on the joint spectral radius”

<sup>7</sup>Raphael Junger, *Lecture Notes in Control and Information Sciences*, 2009  
“The Joint Spectral Radius: Theory and Applications”

# Joint Spectral Radius

- ▶ The problem of determining if the joint spectral radius is less than 1 is undecidable<sup>8</sup> even for “simple” set of matrices  $\Sigma$
- ▶ But lower and upper bounds  $\{\rho_\ell(\Sigma), \rho_u(\Sigma)\}$  can be found via many<sup>9</sup> different analytical methods
- ▶ So if  $\rho_u(\Sigma) < 1$  the stability of the system with (constrained) deadline misses is guaranteed

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<sup>8</sup>Vincent Blondel and John Tsitsiklis, *Systems & Control Letters*, 41(2):135–140, 2000  
“The boundedness of all products of a pair of matrices is undecidable”

<sup>9</sup>Guillaume Vankeerberghen, Julien Hendrickx, and Raphaël M. Jungers, *HSCC 2014*  
“JSR: a toolbox to compute the joint spectral radius”

# Fault Models

- ▶ Probabilistic
- ▶ Constrained, or *weakly-hard*<sup>10</sup>

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<sup>10</sup>Guillem Bernat, Alan Burns, Albert Liamsó, IEEE Transactions on Computers, 2001, “Weakly hard real-time systems”

## Fault Models

- ▶ Probabilistic
- ▶ Constrained, or *weakly-hard*<sup>10</sup>
  1.  $\tau \vdash \binom{x}{k}, \text{AnyHit}$
  2.  $\tau \vdash \langle \frac{x}{k} \rangle, \text{RowHit}$
  3.  $\tau \vdash \overline{\binom{x}{k}}, \text{AnyMiss}$
  4.  $\tau \vdash \overline{\langle \frac{x}{k} \rangle} = \overline{\langle x \rangle}, \text{RowMiss}$

with  $x \in \mathbb{N}^{\geq}, k \in \mathbb{N}^{\>}$ , where  $x \leq k$

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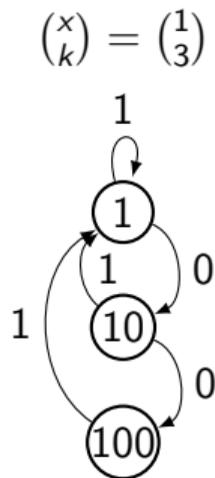
$$\dots 0 \underbrace{10111}_{k} 0110 \dots$$

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# Weakly-Hard Constraints as Automata

- ▶ Any weakly-hard constraint can be transformed into a corresponding finite state machine<sup>a</sup>
- ▶ The transformation enables the analysis via joint spectral radius<sup>b</sup>

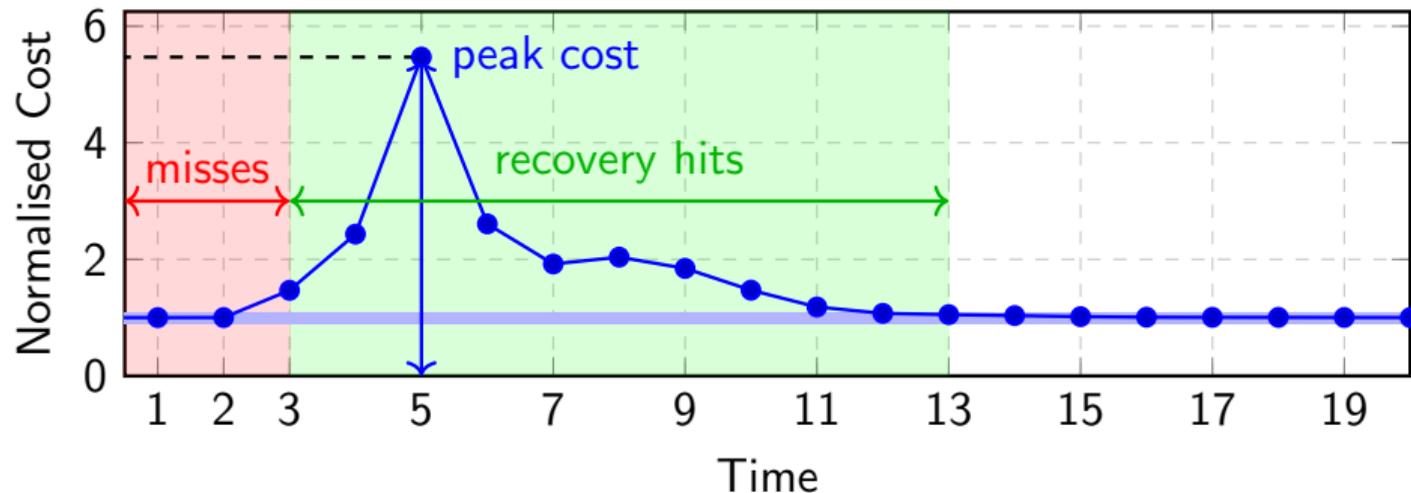


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<sup>a</sup>Nils Vreman, Richard Pates, and Martina Maggio, RTAS 2022  
“WeaklyHard.jl: Scalable Analysis of Weakly-Hard Constraints”  
<https://github.com/NilsVreman/WeaklyHard.jl>

<sup>b</sup>Nils Vreman, Paolo Pazzaglia, Victor Magron, Jie Wang, Martina Maggio, CDC & Letters 2022  
“Stability of Linear Systems Under Extended Weakly-Hard Constraints”

# Performance Analysis<sup>11</sup>



<sup>11</sup>Nils Vreman, Anton Cervin and Martina Maggio, ECRTS 2021  
“Stability and Performance Analysis of Control Systems Subject to Bursts of Deadline Misses”

## Conclusion

- ▶ Stability and performance analysis of control systems subject to deadline misses
- ▶ Sometimes when control software experiences faults (missing deadlines) there is no need to worry!

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